Implementation of a Voltage-Mode All-Pass Filter Using CCII and Its Inverse Configuration

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Abstract: This work presents a voltage-mode (VM) filter capable of performing a first-order all-pass (AP) function as well as generating its inverse signal. The proposed design features a single input and output and utilizes one CCII, two resistors, and a grounded capacitor. By employing either an inverting or non-inverting CCII, the circuit can achieve both AP and inverse AP responses without altering the circuit topology. The phase angle can be adjusted by varying the frequency of the input signal or by modifying the grounded capacitor, maintaining the circuit's reliability condition. Simulation results using the PSPICE software confirm the circuit's performance.

Keywords: Voltage mode Circuit, All Pass Filter, Inverse Function, CCII

I. Introduction

The application of inverse filters is indispensable for recovering signals that have been distorted by a processing or transmission system. This recovery is achievable using an inverse filter block, whose frequency response is the reciprocal of the system that caused the distortion. Therefore, inverse filters are essential components in the design of signal processing and conditioning systems, such as those used in communication, control, and instrumentation. First- and higher-order all-pass filters serve several key functions, including: (1) shifting the phase of a signal from 0 to π while maintaining a constant amplitude over the target frequency range; (2) enabling various filtering characteristics and oscillator designs; and (3) facilitating the realization of high-Q frequencyselective circuits. Numerous circuits implementing all-pass functions in either current-mode or voltage-mode have been reported, utilizing components such as OTAs, FTFNs, or CCIIs [1-9]. However, no existing topology appears to offer the capability to realize both the all-pass function and its inverse simply by reversing the polarity of the active device, without requiring modifications to the circuit structure. To address this, we propose a voltage-mode (VM) filter that achieves both a first-order all-pass (AP) response and its inverse by reversing the polarity of the active device. The proposed design requires only a single CCII, two resistors, and a grounded capacitor. By employing either an inverting or non-inverting CCII, the circuit provides the AP function and its inverse signal, respectively. In both configurations, the phase can be adjusted by varying the input signal frequency or the grounded capacitor, an approach well-suited to integrated circuit (IC) design.

II. Proposed Design

The ideal CCII can be characterized by the following port relations: $V_x = V_y$, $I_y = 0$ and $I_z = \pm Ix$

where the \pm sign depicts the polarity of the CCII. By using non-inverting CCII, the analysis of the circuit in Fig. 1 yields the first-order inverse AP VM transfer function given by:

 $V_0/V_{\text{IN}} = (1/1 \cdot \text{sRC})/1 + \text{sRC}$ ….(1)

By reversing the polarity of the CCII, Eqn 1 takes the form of first order AP signal as given by

 $V_0/V_{IN} = (1 - sRC)/(1 + sRC)$ (2)

The realizibility condition for both the versions of the circuit is $R_1 = R_2$, which is simple and temperature invariant being resistor ratio and therefore desirable in IC fabrication.

The circuit respectively yields phase shifts from 0 to 180 and 0 to -180 for inverse AP function and corresponding AP signal. For an ideal case, the filter

respectively has the following phase responses corresponding to inverse AP and AP:

 $\phi(\omega, R, C) = 2 \arctan(\omega RC)$ (3) $\phi(\omega, R, C) = -2 \arctan(\omega RC)$ (4)

Eqs (3) and (4) reveal that the phase can be controlled by adjusting the frequency of the input signal and/or \overline{C} without influencing the realizibility condition.

Figure1. Proposed voltage-mode filter

III. Error Analysis

Tracking—The non-ideal port relations of CCII are given by $I_z = \alpha I_x$, where $\alpha = 1 - \psi_1$; $\psi_1 \ll 1$, denotes the current tracking error; $V_x =$

 βV_y , where $\beta = 1 - \psi_2$, $\psi_2 \ll 1$, denotes the input voltage tracking error.

The re-analysis of the circuit based on the non- idealities of the device employed yield the following voltagetransfer functions:

 $V_0/V_{\text{IN}} = 1/[(1 - \alpha \beta \text{SRC})/(1 + \text{SRC})]$...(5) $V_0/V_{IN} = (1 - \alpha \beta s RC) / (1 + \alpha \beta s RC)$...(6)

IV. Sensitivity

The study of sensitivities forms an important index of the performance of any active network. The formal definition of sensitivity is:

$Sr_{x}= X\delta F/F\delta X$

where F is the networks function and x is the element of variation. Using this definition, the sensitivities of ω_0 with respect to active and passive components are given by:

S^{w0} α, β, R, C = - 1

which are no more than unity.

V. Experimental Results

PSPICE simulations were performed to verify the feasibility of the circuit yielding responses represented by Eqs (1) and (2). The negative CCII shown in Fig. 2 can be implemented14 by using two AD844s supported by Analog Devices, Inc. The circuit in Fig.1 was built with C=1nF and $R_1=R_2=1K\Omega$ for realizing allpass and its inverse function with a phase shift of 90° at f₀=159KHz

Figs 3 and 4 depict the magnitude responses and the phase response for the two filtering functions, respectively. The phase simulation results obtained agree with the theoretical calculations. However, owing to the non-ideal behaviour of the active device, the magnitude response varies with the theoretical calculations.

Figure 2. Implementation of inverting CCII using two AD844

Figure 3. Magnitude responses of AP and inverse AP functions

Figure 4. Phase response of AP and inverse AP functions

VI. Conclusions

A novel generic circuit implementing first-order allpass and its inverse function has been presented. The circuit is based on low component count and has simple realizibility condition being in the ratio of resistors, which remains insensitive to environmental changes. By using inverting CCII, the circuit permits implementation of AP function while inverse AP function can be realized by changing the polarity of the device. Both the filtering signals have been implemented without inducing change in the circuit topology or affecting rotation of components or using additional components.

VII. References

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